

## SYNCHRONIZATION CIRCUIT

FILED OF THE INVENTION

The present invention relates to a synchronization circuit for synchronizing an asynchronously inputted signal with a clock, in a digital signal transmission apparatus.

BACKGROUND OF THE INVENTION

A conventional synchronization circuit synchronizes an asynchronously inputted signal with a synchronization clock, and outputs the synchronized signal (refer to Japanese Published Patent Application No. 5-327676 and USP4965814). Hereinafter, the conventional synchronization circuit will be described with reference to figure 17.

Figure 17 is a block diagram illustrating the construction of the conventional synchronization circuit.

With reference to figure 17, a flip-flop 1 receives an input signal SIN that is asynchronous to a synchronization clock SCK, and an inverse clock nSCK that is output from an inverter 5, and latches the input signal SIN at a timing of a rising edge of the inverse clock nSCK. A flip-flop 2 receives the input signal SIN and the synchronization clock SCK, and latches the input signal SIN at a timing of a rising edge of the synchronization clock SCK. A flip-flop 3 receives a signal selected by a selection circuit 4 and the synchronization clock SCK, and outputs a synchronizing signal SOUT at a timing of the rising edge of the synchronization clock SCK. The selection circuit 4 selects either the output of

the flip-flop 1 or the output of the flip-flop 2 on the basis of a control signal CTL that is output from a switching control circuit 6. The inverter 5 receives the synchronization clock SCK, and outputs an inverse clock nSCK that is obtained by inverting the synchronization clock SCK. The switching control circuit 6 outputs a control signal CTL according to the temporal relationship between a transition point of the input signal SIN and an edge of the synchronization clock SCK.

Hereinafter, the operation of the synchronization circuit constructed as described above will be described.

The asynchronous input signal SIN is applied to data terminals D of the flip-flops 1 and 2.

When the inverse clock nSCK outputted from the inverter 5 is input to the flip-flop 1 through a clock input terminal CK, the flip-flop 1 latches the input signal SIN at a timing of the rising edge of the inverse clock nSCK, and outputs the signal to the selection circuit 4 through a data output terminal Q.

Further, when the synchronization clock SCK is input to the flip-flop 2 through a clock input terminal CK, the flip-flop 2 latches the input signal SIN at a timing of the rising edge of the synchronization clock SCK, and outputs the signal to the selection circuit 4 through a data output terminal Q.

On the other hand, the switching control circuit 6 monitors the temporal relationship between the transition point of the asynchronous input signal SIN and the edge of the synchronization

clock SCK, and outputs the control signal CTL when detecting that the relationship approaches a predetermined period of time, thereby to control the selection circuit 4.

The flip-flop 3 latches the signal selected by the selection circuit 4 at a timing of the rising edge of the synchronization clock SCK, and outputs a synchronizing signal SOUT through a data output terminal Q.

In this way, the asynchronous input signal SIN is synchronized with the synchronization clock SCK.

However, the signal latched at the inverse clock nSCK has already been output from the selection circuit 4 when the switching control circuit 6 detects that the transition point of the asynchronous input signal SIN approaches the edge of the synchronous clock SCK, and this signal is again latched at the synchronization clock SCK by the third flip-flop 3, and thereby a latency is undesirably added to the signal.

Furthermore, there are many cases where plural pieces of asynchronous signals are input in recent multi-channel digital transmission, and skews between the plural input signals adversely affect data transmission as the input signals become faster. Since, in the conventional technique, there is a possibility that a latency is added to every signal, such skews cause a serious problem in data transmission in which error-free signal synchronization should be carried out.

#### SUMMARY OF THE INVENTION

The present invention is made to solve the above-described problems and has for its object to provide a synchronization circuit that can prevent addition of latency to an input signal, and is reduced in circuit scale.

Other objects and advantages of the invention will become apparent from the detailed description that follows. The detailed description and specific embodiments described are provided only for illustration since various additions and modifications within the scope of the invention will be apparent to those of skill in the art from the detailed description.

According to a first aspect of the present invention, there is provided a synchronization circuit for receiving an input signal and a clock having a frequency equal to a transfer rate of the input signal, and synchronizing the input signal with the clock, and the synchronization circuit comprises a state detection circuit for outputting a control signal according to the temporal relationship between a transition point of the input signal and an edge of the clock; a delay selection circuit for adding a delay to the input signal on the basis of the control signal; and a latch circuit for synchronizing the signal outputted from the delay selection circuit with the clock, and outputting the synchronized signal. Therefore, the input signal can be synchronized with the inputted clock without the necessity for inverting the input signal as in the conventional circuit. As a result, a synchronization circuit that can perform the

above-described synchronization without adding latency to the input signal can be implemented in a relatively simple construction.

According to a second aspect of the present invention, there is provided a synchronization circuit for receiving an input signal and a clock having a frequency equal to a transfer rate of the input signal, and synchronizing the input signal with the clock, and the synchronization circuit comprises a state detection circuit for outputting a control signal according to the temporal relationship between a transition point of the input signal and an edge of the clock; a delay selection circuit for adding a delay to the clock on the basis of the control signal; and a latch circuit for synchronizing the input signal with the clock outputted from the delay selection circuit, and outputting the synchronized signal. Therefore, the input signal can be synchronized with the inputted clock without the necessity for inverting the input signal as in the conventional circuit. As a result, a synchronization circuit that can perform the above-described synchronization without adding latency to the input signal can be implemented in a relatively simple construction.

According to a third aspect of the present invention, there is provided a synchronization circuit for receiving plural input signals having phases irrelevant to each other, and a clock having a frequency equal to a transfer rate of the plural input signals, and synchronizing the plural input signals with the

clock, and the synchronization circuit comprises a state detection circuit for outputting control signals relating to the respective input signals, according to the temporal relationship between transition points of the plural input signals; a delay selection circuit for adding delays to the respective input signals on the basis of the control signals relating to the respective input signals; and a latch circuit for synchronizing the respective signals outputted from the delay selection circuit with the clock, and outputting the synchronized signals.

Therefore, each input signal can be synchronized with the inputted clock without the necessity for inverting the input signal as in the conventional circuit. As a result, a synchronization circuit that can perform the above-described synchronization without adding latency to the input signal can be implemented in a relatively simple construction.

According to a fourth aspect of the present invention, there is provided a synchronization circuit for receiving plural signal bundles each comprising a set of plural input signals synchronized with each other and a single clock having a frequency equal to a transfer rate of the plural input signals, in which the phases of the input signals included in one signal bundle are irrelevant to the phases of the input signals included in the other signal bundles, and synchronizing the input signals included in one signal bundle with the input signals included in the other signal bundles by using a single synchronization clock

that is selected from among the clocks included in the respective signal bundles, and the synchronization circuit comprises a state detection circuit for detecting the state between the plural input signals included in the respective signal bundles; a clock selection circuit for receiving the clocks included in the respective signal bundles, and selecting one of the inputted clocks, as a synchronization clock, on the basis of the result of the state detection performed between the respective signal bundles by the state detection circuit; a delay selection circuit for adding delays to the plural input signals included in each signal bundle, on the basis of the result of the state detection performed between the respective signal bundles; and a latch circuit for synchronizing the output signal from the delay selection circuit for each signal bundle, with the synchronization clock, and outputting the synchronized signal. Therefore, the plural signal bundles which are inputted asynchronously with each other can be synchronized with each other without inverting the plural input signals included in the respective signal bundles. As a result, a synchronization circuit that can perform synchronization without adding latency to the input signals can be implemented in a relatively simple construction.

According to a fifth aspect of the present invention, in the synchronization circuit according to the fourth aspect, the state detection circuit comprises an early/late detection circuit for

detecting which signal bundle is earlier in input timing between the respective signal bundles, and outputting an early/late detection signal; and an overlap detection circuit for detecting an overlap period between the respective signal bundles, and outputting an overlap detection signal; the clock selection circuit selects, as a synchronization clock, a clock included in a signal bundle which is determined as being inputted earlier between the respective signal bundles, on the basis of the early/late detection signal; and the delay selection circuit adds delays based on the early/late detection signal and the overlap detection signal, to the plural input signals included in the respective signal bundles. Therefore, a synchronization circuit that can synchronize plural signal bundles regardless of the presence or absence of an overlap period between data to be synchronized, which data are included in the respective signal bundles, can be implemented in a relatively simple construction.

According to a sixth aspect of the present invention, in the synchronization circuit according to the first aspect, the delay selection circuit comprises a delay circuit for adding a delay to the input signal; and a selection circuit for selecting either the input signal or the output signal of the delay circuit on the basis of the control signal. Since a delay is added to the input signal on the basis of the control signal, it becomes unnecessary to invert the input signal as in the conventional circuit, resulting in a synchronization circuit that can avoid addition of



latency to the input signal.

According to a seventh aspect of the present invention, in the synchronization circuit according to the second aspect, the delay selection circuit comprises a delay circuit for adding a delay to the inputted clock; and a selection circuit for selecting either the inputted clock or the clock outputted from the delay circuit on the basis of the control signal. Therefore, a synchronization clock, which is obtained by adding a delay to the clock on the basis of the control signal, can be used for synchronizing the input signal. As a result, it becomes unnecessary to invert the input signal as in the conventional circuit, resulting in a synchronization circuit that can avoid addition of latency to the input signal.

According to an eighth aspect of the present invention, in the synchronization circuit according to the third aspect, the delay selection circuit comprises a delay circuit for adding delays to the respective input signals; and a selection circuit for selecting one from among the plural input signals and the signals outputted from the delay circuit, for each of the plural input signals, on the basis of the control signals relating to the respective input signals, and outputting the selected signal. Since a delay is added to each input signal on the basis of each control signal, it becomes unnecessary to invert the input signal as in the conventional circuit, resulting in a synchronization circuit that can avoid addition of latency to the input signal.

According to a ninth aspect of the present invention, in the synchronization circuit according to any of the first to fifth aspects, the state detection circuit detects the state of the input signal on the basis of a preamble detection signal which is supplied from the outside and indicates the positional relationship of data to be synchronized. Therefore, the positional relationship of the data to be synchronized can be easily determined.

#### BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a waveform diagram of an input signal.

Figure 2 is a block diagram illustrating the construction of a synchronization circuit according to a first embodiment.

Figure 3 is a timing chart in the synchronization circuit according to the first embodiment.

Figure 4 is a timing chart in the synchronization circuit according to the first embodiment.

Figure 5 is a diagram illustrating the construction of a state detection circuit.

Figure 6 is a block diagram illustrating the construction of a synchronization circuit according to a second embodiment of the present invention.

Figure 7 is a timing chart in the synchronization circuit according to the second embodiment.

Figure 8 is a timing chart in the synchronization circuit according to the second embodiment.

Figure 9 is a block diagram illustrating the construction of a synchronization circuit according to a third embodiment of the present invention.

Figure 10 is a block diagram illustrating the construction of a synchronization circuit according to a fourth embodiment of the present invention.

Figure 11 is a timing chart illustrating input signals to be input to the synchronization circuit according to the fourth embodiment.

Figure 12 is a diagram illustrating the construction of a state detection circuit included in the synchronization circuit according to the fourth embodiment.

Figure 13 is a diagram illustrating the construction of an early/late detection circuit included in the synchronization circuit according to the fourth embodiment.

Figure 14 is a diagram illustrating the construction of an overlap detection circuit included in the synchronization circuit according to the fourth embodiment.

Figure 15 a diagram illustrating the construction of a delay selection circuit included in the synchronization circuit according to the fourth embodiment.

Figure 16 is a timing chart illustrating input signals to be input to the synchronization circuit according to the fourth embodiment.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Hereinafter, embodiments of the present invention will be described with reference to the drawings. The embodiments described below are merely examples, and the present invention is not restricted thereto.

[Embodiment 1]

Hereinafter, a synchronization circuit according to a first embodiment of the present invention will be described with reference to figures 1 and 2.

Figure 1 is a waveform diagram illustrating an asynchronous input signal SIN that is input to a synchronization circuit according to the present invention, wherein a period  $T_s$  is a signal definite period in which a set-up hold time of the signal SIN is ensured, and a period  $T_d$  is a signal indefinite period in the vicinity of a transition point of the signal SIN.

Figure 2 is a block diagram illustrating the construction of the synchronization circuit according to the first embodiment.

The synchronization circuit shown in figure 2 is provided with a state detection circuit 102 for outputting a control signal CTL according to the temporal relationship between a transition point of the input signal SIN and an edge of a synchronization clock SCK, a delay selection circuit 101 for adding a delay to the input signal SIN on the basis of the control signal CTL, and a latch circuit (flip-flop) 103 for synchronizing an output signal SD of the delay selection circuit 101 with the synchronization clock SCK.

The state detection circuit 102 sets the control signal CTL at "High" and outputs it, when the edge of the synchronization clock SCK exists in a period where a sufficient set-up hold time is not ensured for the input signal SIN, i.e., in the signal indefinite period  $T_d$  shown in figure 1. On the other hand, the state detection circuit 102 sets the control signal CTL at "Low" and does not output it, when the edge of the synchronization clock SCK exists in the signal definite period  $T_s$ . As shown in figure 5, the state detection circuit 102 can be implemented as a circuit comprising a delay circuit 104 which adds a delay to the input signal SIN, an XOR circuit 105 which receives the input signal SIN and an output signal  $DS_i$  of the delay circuit 104, and outputs a signal  $S_{xor}$ , and a flip-flop 106 which receives the output signal  $S_{xor}$  of the XOR circuit 105 and the synchronization clock SCK, and outputs the control signal CTL.

The delay selection circuit 101 comprises a delay circuit 111 for adding a delay to the input signal SIN, and a selection circuit (2:1 selector) 112 for selecting either the input signal SIN or an output signal  $DSIN$  of the delay circuit 111 on the basis of the control signal CTL outputted from the state detection circuit 102. The selection circuit 112 selects the output signal  $DSIN$  of the delay circuit 111 when the control signal CTL is input thereto.

Hereinafter, the operation of the synchronization circuit constructed as described above will be described with reference

to figures 3 and 4.

The input signal SIN is input to the delay selection circuit 101 and the state detection circuit 102, and the synchronization clock SCK is input to the state detection circuit 102 and the flip-flop 103. While the transfer rate of the input signal SIN is equal to the frequency of the synchronization clock SCK, the phase of the input signal SIN is irrelevant to the phase of the synchronization clock SCK.

Initially, in the delay selection circuit 101, the delay circuit 111 adds a delay to the input signal SIN, and outputs the delay-added signal DSIN to the selection circuit 112.

On the other hand, the state detection circuit 102 performs a comparison of phases between the input clock CK and the input signal SIN.

As a result of the phase comparison, when it is detected that an edge e11 of the synchronization clock SCK exists in the signal definite period  $T_s$  of data d11 of the input signal SIN as shown in figure 3, the control signal CTL remains at "Low" and is not output to the delay selection circuit 101. Accordingly, in the selection circuit 112 in the delay selection circuit 101, the input signal SIN is selected and outputted as a signal SD to the flip-flop 103. In the flip-flop 103, the data d11 of the signal SD outputted from the delay selection circuit 101 is synchronized with the synchronization clock SCK to be output as a synchronizing signal SOUT.

Further, as a result of the phase comparison, when it is detected that an edge e12 of the synchronization clock SCK exists in the signal indefinite period Td of data d12 of the input signal SIN as shown in figure 4, the control signal CTL is changed from "Low" to "High" to be output to the delay selection circuit 101. Accordingly, in the selection circuit 112 in the delay selection circuit 101, the output signal of the delay circuit 111, i.e., the signal DSIN obtained by adding a delay time Tdel to the input signal SIN, is selected and outputted as a signal SD to the flip-flop 103. In the flip-flop 103, the data d12 of the signal SD outputted from the delay selection circuit 101 is latched at an edge e13 of the synchronization clock SCK to be output as a synchronizing signal SOUT.

While in the above description the operation of the selection circuit 112 is switched when the output from the state detection circuit 102 is "High", the present invention is not restricted thereto.

The synchronization circuit according to the first embodiment is provided with the state detection circuit 102 for outputting the control signal CTL according to the temporal relationship between the transition point of the input signal SIN and the edge of the synchronization clock SCK, the delay selection circuit 101 for adding a delay to the input signal SIN on the basis of the control signal CTL, and the latch circuit 103 for synchronizing the signal SD outputted from the delay

selection circuit 101 with the synchronization clock SCK. Since it is not necessary to invert the input signal as in the conventional circuit, the input signal SIN can be synchronized with the synchronization clock SCK without considering the temporal relationship between the signal indefinite period of the input signal SIN and the edge of the synchronization clock SCK. As a result, a synchronization circuit that can perform the above-described synchronization without adding latency to the input signal can be implemented in a relatively simple construction.

[Embodiment 2]

Hereinafter, a synchronization circuit according to a second embodiment of the present invention will be described with reference to figure 6.

Figure 6 is a block diagram illustrating the construction of the synchronization circuit according to the second embodiment.

The synchronization circuit shown in figure 6 is provided with a state detection circuit 202 for outputting a control signal CTL according to the temporal relationship between a transition point of an input signal SIN and an edge of an input clock CK, a delay selection circuit 201 for adding a delay to the input clock CK on the basis of the control signal CTL, and a flip-flop 203 for synchronizing the input signal SIN with a clock SCK that is selected by the delay selection circuit 201.

The state detection circuit 202 sets the control signal CTL



at "High" and outputs it, when the edge of the synchronization clock SCK exists within a period in which a sufficient set-up hold time is not ensured for the input signal SIN, i.e., the signal indefinite period  $T_d$  shown in figure 1. On the other hand, the state detection circuit 202 sets the control signal CTL at "Low" and does not output it, when the edge of the synchronization clock SCK exists within the signal definite period  $T_s$  shown in figure 1. The state detection circuit 202 can be implemented by a circuit shown in figure 5.

The delay selection circuit 201 is provided with a delay circuit 211 for adding a delay to the input clock CK, and a selection circuit (2:1 selector) 212 for selecting either the input clock CK or an output clock DCK of the delay circuit 211 on the basis of the control signal CTL. The selection circuit 212 selects the output clock DCK of the delay circuit 211 when the control signal CTL is input thereto.

The operation of the synchronization circuit constructed as described above will be described.

The input signal SIN is input to the state detection circuit 202 and the flip-flop 203, and the input clock CK is input to the state detection circuit 202 and the delay selection circuit 201. Although the transfer rate of the input signal SIN is equal to the frequency of the input clock SCK, the phase of the input signal SIN is irrelevant to the phase of the input clock SCK.

First of all, in the delay selection circuit 201, the delay

circuit 211 adds a delay to the input clock CK, and outputs the delay-added clock DCK to the selection circuit 212.

On the other hand, the state detection circuit 202 performs a comparison of phases between the input clock CK and the input signal SIN.

As a result of the phase comparison, when it is detected that an edge e21 of the input clock exists in the signal definite period  $T_s$  of data d21 of the input signal SIN as shown in figure 7, the control signal CTL remains at "Low" and is not output to the delay selection circuit 201. Accordingly, in the selection circuit 212 in the delay selection circuit 201, the input clock CK is selected and outputted as a synchronization clock SCK to the flip-flop 203. In the flip-flop 203, the data d21 of the input signal SIN is synchronized with the synchronization clock SCK to be output as a synchronizing signal SOUT.

Further, as a result of the phase comparison, when it is detected that an edge e22 of the input clock exists in the signal indefinite period  $T_d$  of data d22 of the input signal SIN as shown in figure 8, the control signal CTL is changed from "Low" to "High" to be output to the delay selection circuit 201. Accordingly, in the selection circuit 212 in the delay selection circuit 201, the output signal of the delay circuit 211, i.e., a clock DCK obtained by adding a delay time  $T_{del}$  to the input clock CK, is selected and outputted as a synchronization clock CK to the flip-flop 203. In the flip-flop 203, the data d22 of the

input signal SIN is latched at an edge e23 of the clock SCK outputted from the delay selection circuit 201 to be output as a synchronizing signal SOUT.

While in the above description the operation of the selection circuit 212 is switched when the output from the state detection circuit 202 is "High", the present invention is not necessarily restricted thereto.

The synchronization circuit according to the second embodiment is provided with the state detection circuit 202 for outputting the control signal CTL according to the temporal relationship between the transition point of the input signal SIN and the edge of the clock CK, the delay selection circuit 201 for adding a delay to the clock CK on the basis of the control signal CTL, and the latch circuit 203 for synchronizing the input signal SIN with the clock SCK outputted from the delay selection circuit 201. Since it is not necessary to invert the input signal as in the conventional circuit, the input signal SIN can be synchronized with the synchronization clock SCK without considering the temporal relationship between the signal indefinite period of the input signal SIN and the edge of the synchronization clock SCK. As a result, a synchronization circuit that can perform the above-mentioned synchronization without adding latency to the input signal can be implemented in a relatively simple construction.

[Embodiment 3]

Hereinafter, a synchronization circuit according to a third embodiment of the present invention will be described with reference to figure 9.

Figure 9 is a block diagram illustrating the construction of the synchronization circuit according to the third embodiment.

The synchronization circuit shown in figure 9 is provided with a state detection circuit 303 for outputting a first control signal CTL1 and a second control signal CTL2 according to the temporal relationships between transition points of a first input signal SIN1 and a second input signal SIN2, and an edge of a synchronization clock SCK, respectively; a delay selection circuit 301 for adding a delay to the first input signal SIN1 on the basis of the first control signal CTL1; a delay selection circuit 302 for adding a delay to the second input signal SIN2 on the basis of the second control signal CTL2; a flip-flop 304 for synchronizing the output signal of the delay selection circuit 301 with the synchronization clock SCK to output a first synchronizing signal SOUT; and a flip-flop 305 for synchronizing the output signal of the delay selection circuit 302 with the synchronization clock SCK to output a second synchronizing signal SOUT2.

The state detection circuit 303 outputs the control signal CTL when the edge of the synchronization clock SCK exists within a period in which a sufficient set-up hold time is not secured for the input signal, i.e., the signal indefinite period  $T_d$  shown

in figure 1, and it doesn't output the control signal CTL when the edge of the synchronization clock SCK exists within the signal definite period  $T_s$  shown in figure 1. The state detection circuit 303 can be implemented by the circuit shown in figure 5.

Further, the delay selection circuit 301 is provided with the delay circuit 311 for adding a delay to the first input signal SIN1; and a selection circuit (2:1 selector) 312 for selecting either the first input signal SIN1 or the output signal DSIN of the delay circuit 311 on the basis of the first control signal CTL1. The selection circuit 312 selects the output signal SDIN of the delay circuit 311 when the first control signal CTL1 is input thereto.

Further, the delay selection circuit 302 is provided with a delay circuit 321 for adding a delay to the second input signal SIN2; and a selection circuit (2:1 selector) 322 for selecting either the second input signal SIN2 or the output signal DSIN2 of the delay circuit 321 on the basis of the second control signal CTL2. The selection circuit 322 selects the output signal DSIN2 of the delay circuit 321 when the second control signal CTL2 is input thereto.

Hereinafter, the operation of the synchronization circuit constructed as described above will be described.

The first input signal SIN1 is input to the delay selection circuit 301 and the state detection circuit 303, the second input signal SIN2 is input to the delay selection circuit 302 and the

state detection circuit 303, and the synchronization clock SCK is input to the state detection circuit 303, the flip-flop 304, and the flip-flop 305. While the transfer rates of the respective input signals SIN1 and SIN2 are equal to the frequency of the synchronization clock SCK, the phases of the input signals SIN1 and SIN2 are irrelevant to the phase of the synchronization clock SCK.

Initially, in the delay selection circuit 301, the delay circuit 311 adds a delay to the first input signal SIN1, and outputs the delay-added signal DSIN to the selection circuit 312. Further, in the delay selection circuit 302, the delay circuit 321 adds a delay to the second input signal SIN2, and outputs the delay-added signal DSIN to the selection circuit 322.

On the other hand, the state detection circuit 303 performs a comparison of phases between the synchronization clock SCK and the input signals SIN1 and SIN2.

As a result of the phase comparison, when it is detected that the edge of the synchronization clock SCK exists in the signal indefinite period  $T_d$  of the first input signal SIN1, the first control signal CTL1 is output to the delay selection circuit 301. Accordingly, in the selection circuit 312 in the delay selection circuit 301, the output signal DSIN1 of the delay circuit 311 is selected and outputted to the flip-flop 304 as a signal SD1.

Further, when it is detected that the edge of the

synchronization clock SCK exists in the signal definite period  $T_s$  of the first input signal SIN1, the first control signal CTL1 is not output to the delay selection circuit 301 and, therefore, the first input signal SIN1 is selected by the selection circuit 312 and outputted to the flip-flop 304 as a signal SD1.

Further, when it is detected that the edge of the synchronization clock SCK exists in the signal indefinite period  $T_d$  of the second input signal SIN2, the second control signal CTL2 is output to the delay selection circuit 302. Accordingly, in the selection circuit 322 in the delay selection circuit 302, the output signal DSIN2 of the delay circuit 321 is selected and outputted to the second flip-flop 305 as a signal SD2.

Further, when it is detected that the edge of the synchronization clock SCK exists in the signal definite period  $T_s$  of the second input signal SIN2, the second control signal CTL2 is not output to the delay selection circuit 302 and, therefore, the second input signal SIN2 is selected by the selection circuit 322 and outputted to the flip-flop 305 as a signal SD2.

In the flip-flop 304, the output signal SD from the delay selection circuit 301 is latched at the edge of the synchronization clock SCK to be output as a synchronizing signal SOUT1. Further, in the flip-flop 305, the output signal SD from the delay selection circuit 302 is latched at the edge of the synchronization clock SCK to be output as a synchronizing signal SOUT2.

While in the above-description two input signals are adopted, an arbitrary number of input signals (not less than two) may be adopted. At this time, the number of times of state detection in the state detection circuit 303 changes according to the number of input signals.

The synchronization circuit according to the third embodiment is provided with the state detection circuit 303 for outputting the control signals CTL1 and CTL2 relating to the respective input signals SIN1 and SIN2, according to the temporal relationship between the transition points of the input signals SIN1 and SIN2; the delay selection circuit 302 for adding delays to the respective input signals SIN1 and SIN2 on the basis of the control signals CTL1 and CTL2; and the latch circuits 304 and 305 for synchronizing the signals SD1 and SD2 outputted from the delay selection circuit 302 with the synchronization clock SCK. Since it is not necessary to invert each input signal as in the conventional circuit, the first input signal SIN1 and the second input signal SIN2 can be synchronized with each other using the synchronization clock SCK without considering the temporal relationship between the signal indefinite period of each input signal and the edge of the synchronization clock SCK. As a result, a synchronization circuit that can perform the above-described synchronization without adding latency to the input signals can be implemented in a relatively simple construction.

[Embodiment 4]



Hereinafter, a synchronization circuit according to a fourth embodiment of the present invention will be described with reference to figure 10.

The synchronization circuit according to the fourth embodiment receives plural signal bundles each comprising a set of plural input signals synchronized with each other and a single clock having a frequency equal to a transfer rate of the plural input signals, in which the phases of the input signals included in one signal bundle are irrelevant to the phases of the input signals included in the other signal bundles, and synchronizes the input signals included in one signal bundle with the input signals included in the other signal bundles by using a single synchronization clock that is selected from among the clocks included in the respective signal bundles.

Figure 10 is a block diagram illustrating the construction of the synchronization circuit according to the fourth embodiment of the present invention. While plural signal bundles should be input to the synchronization circuit, figure 10 shows two signal bundles each comprising a set of a single input signal and a single clock for the sake of convenience. To be specific, in figure 10, SIN-1 denotes one of signals included in the first signal bundle, SIN-2 denotes one of signals included in the second signal bundle, CK1 is a clock included in the first signal bundle, and CK2 is a clock included in the second signal bundle.

The synchronization circuit shown in figure 10 is provided

with a state detection circuit 401 for detecting the state between the first input signal SIN-1 and the second input signal SIN-2 which are included in the respective signal bundles; a clock selection circuit 402 for selecting either the first input clock CK1 or the second input clock CK2 on the basis of a result of the state detection performed between the respective signal bundles by the state detection circuit 401, and outputting the selected clock as a synchronization clock SCK; a delay selection circuit 403 for adding a delay to the first input signal SIN-1 on the basis of the result of the state detection performed between the respective signal bundles, and outputting the delayed signal as a signal SD11; a delay selection circuit 404 for adding a delay to the second input signal SIN-2 on the basis of the result of the state detection performed between the respective signal bundles, and outputting the delayed signal as a signal SD21a; a flip-flop 405 for synchronizing the signal SD11 with the synchronization clock SCK; and a flip-flop 406 for synchronizing the signal SD21 with the synchronization clock SCK.

As shown in figure 12, the state detection circuit 401 is provided with an early/late detection circuit 407 for detecting as to which signal bundle is earlier in input timing between the respective signal bundles (SIN-1 and SIN-2), and an overlap detection circuit 408 for detecting an overlap period between the respective signal bundles (SIN-1 and SIN-2).

As shown in figure 13, the early/late detection circuit 407

is provided with a flip-flop 444 for receiving the first input signal SIN-1, and a flip-flop 445 for receiving the second input signal SIN-2. One of the flip-flops 444 and 445, which receives the input signal earlier than the other flip-flop, outputs a signal Kil or Ki2 indicating that inputting should be stopped, to the other flip-flop. When the first input signal SIN-1 is inputted earlier than the second input signal SIN-2, the flip-flop 444 outputs an early/late detection signal Fa1. When the second input signal SIN-2 is inputted earlier than the first input signal SIN-1, the flip-flop 445 outputs an early/late detection signal Fa2.

As shown in figure 14, the overlap detection circuit 408 is provided with a delay circuit 421 for adding a delay to the first input signal SIN-1, a delay circuit 422 for adding a delay to the second input signal SIN-2, an AND circuit 426 for receiving the first input signal SIN-1 and the output signal of the delay circuit 422, an AND circuit 427 for receiving the second input signal SIN-2 and the output signal of the delay circuit 421, an AND circuit 428 for receiving the output signal of the AND circuit 426 and the output signal of the AND circuit 427, an XOR circuit 429 for receiving the output signal of the flip-flop 423 and the output signal of the flip-flop 424, a flip-flop 423 for receiving the output signal of the AND circuit 426, a flip-flop 424 for receiving the output signal of the AND circuit 427, and a flip-flop 428 for receiving the output signal of the AND circuit

428.

As shown in figure 15, the first delay selection circuit 403 is provided with a flip-flop 431 for receiving the first input signal SIN-1 and the first clock signal CK1, and outputting a signal AD; a flip-flop 432 for receiving the output signal AD of the flip-flop 431 and the first clock signal CK1, and outputting a signal BD; a selection circuit 433 for selecting the output signal BD of the flip-flop 432 when either the early/late detection signal Fal or the overlap detection circuit Ov1 is input, while selecting the output signal AD of the flip-flop 431 when no signal is inputted; a delay circuit 435 for adding a delay time Tdelay to the output signal SIN-S1 of the selection circuit 433, and outputting a delay-added signal SIN-D1; and a selection circuit 434 for selecting the output signal SIN-S1 of the selection circuit 433 when the overlap detection signal So1 is not inputted, while selecting the output signal SIN-D1 of the delay circuit 435 when the overlap detection signal So1 is inputted. The selection circuit 433 and the selection circuit 434 are 2:1 selectors.

The operation of the synchronization circuit constructed as described above will be described.

The first input signal SIN-1 is input to the delay selection circuit 403 and the state detection circuit 401, the second input signal SIN-2 is input to the delay selection circuit 404 and the state detection circuit 401, the first input clock CK1 is input

to the delay selection circuit 403 and the clock selection circuit 402, and the second input clock CK2 is input to the delay selection circuit 404 and the clock selection circuit 402. While the first input signal SIN-1 and the first input clock CK1 (the second input signal SIN-2 and the second input clock) are inputted in synchronization with each other as shown in figure 11, the first input signal SIN-1 and the second input signal SIN-2 are asynchronous to each other. Further, data d41 and data d42 are signals to be synchronized.

Initially, in the state detection circuit 401, the early/late detection circuit 407 detects as to which signal is earlier in input timing between the first input signal SIN-1 and the second input signal SIN-2, and outputs a detection signal Fa1 to the clock selection circuit 402 when the first input signal SIN-1 is earlier, or outputs a detection signal Fa2 to the clock selection circuit 402 when the second input signal SIN-2 is earlier. When the detection signal Fa1 is input to the clock selection circuit 402, the clock selection circuit 402 outputs the first input clock CK1 as a synchronization clock SCK to the first flip-flop 405 and the second flip-flop 406. On the other hand, when the detection signal Fa2 is input to the clock selection circuit 402, the clock selection circuit 402 outputs the second input clock CK2 as a synchronization clock SCK to the first flip-flop 405 and the second flip-flop 406.

On the other hand, in the state detection circuit 401, the

overlap detection circuit 408 detects an overlap period between the first input signal SIN-1 and the second input signal SIN-2. When the overlap period is longer than a delay time  $T_{so}$ , the overlap detection circuit 408 outputs a detection signal Ov1 and a detection signal Ov2 to the first delay selection circuit 403 and the second delay selection circuit 404, respectively. When the overlap period is shorter than the delay time  $T_{so}$ , the overlap detection circuit 408 outputs a detection signal So1 and a detection signal So2 to the first delay selection circuit 403 and the second delay selection circuit 404, respectively.

In the first delay selection circuit 403, when either the detection signal Ov1 or the detection signal Fal from the state detection circuit 401 is input to the selection circuit 433, the output signal BD from the flip-flop 432 is selected. Otherwise, the output signal AD from the flip-flop 431 is selected. Further, when the detection signal So1 is input to the selection circuit 434, the output signal SIN-D1 of the delay circuit 435 is selected and outputted as a signal SD11 to the first flip-flop 405. Otherwise, the output signal SIN-S1 of the selection circuit 433 is outputted as a signal SD11. Then, in the first flip-flop 405, the output signal SD11 of the delay selection circuit 403 is synchronized with the synchronization clock SCK outputted from the clock selection circuit 402 to be output as a synchronizing signal SOUT11.

Further, the second delay selection circuit 404 is

controlled by the output signals Ov2, So2, and Fa2 of the state detection circuit 401 in like manner as described for the first delay selection circuit 403, and a signal SD21 is output to the second flip-flop 406. Then, in the second flip-flop 406, the output signal SD21 of the delay selection circuit 404 is synchronized with the synchronization clock SCK outputted from the clock selection circuit 402 to be output as a synchronizing signal SOUT21.

In this way, the first signal bundle SIN-1 and the second signal bundle SIN-2, which have been inputted asynchronously to each other, are synchronized. Since these signals are synchronized even when the data d43 and d44 to be synchronized have no overlap period as shown in figure 16, displacements of pictures or the like can be avoided.

While in this fourth embodiment two input signals are adopted, an arbitrary number of input signals not less than two may be adopted. Further, as for clocks to be input in synchronization with the respective input signals, an arbitrary number of clocks not less than two may be adopted. Thus, the number of input signals and the number of clocks may be arbitrarily selected as long as the above-mentioned functions are satisfied, and the present invention is not restricted to the above-described construction.

The synchronization circuit according to the fourth embodiment is provided with the state detection circuit 401 for

receiving two signal bundles each comprising a set of plural synchronous input signals and a single clock having a frequency equivalent to a transfer rate of the plural input signals, in which the input signals included in one signal bundle is irrelevant to the input signals included in the other signal bundles, and detecting the state between the input signals included in the respective signal bundles; the clock selection circuit 402 for receiving the clock CK1 and the clock CK2 included in the respective signal bundles, and selecting one of the input clocks CK1 and CK2 as a synchronization clock SCK on the basis of the result of the state detection performed by the state detection circuit 401; the delay selection circuits 403 and 404 for adding delays to the plural input signals SIN-1 and SIN-2 included in the respective signal bundles on the basis of the result of the state detection performed between the respective signal bundles; and the latch circuits 405 and 406 for synchronizing the output signals SD11 and DS21 from the delay selection circuits 403 and 404 with the synchronization clock SCK, respectively. Since it is not necessary to invert the input signal SIN-1 and the input signal SIN-2 as in the conventional circuit, the input signals SIN-1 and SIN-2 can be synchronized with each other using the synchronization clock SCK without considering the temporal relationship between the signal indefinite period of each input signal and the edge of the synchronization clock. As a result, a synchronization circuit



that can perform the above-mentioned synchronization of the respective input signals SIN-1 and SIN-2 without adding latency to the input signals even when there is no overlap period of data to be synchronized, can be implemented in a relatively simple construction.

In the respective embodiments of the present invention, a preamble signal indicating the positional relationship between the data to be synchronized may be input to the state detection circuit so as to detect a preamble pattern of the input signal, whereby the positional relationship between the data to be synchronized can easily be detected.

The synchronization circuit according to the present invention is useful as a circuit capable of increasing the data transmission efficiency in a data transmission system such as a digital transmission apparatus.